

LTM2810

7.5kVRMS SPI/Digital or I²C µModule Isolator with Transformer Driver

- 6-Channel Logic Isolator: 7500V_{RMS} for 1 Minute
- ⁿ **16.2mm Creepage**
- ⁿ **CSA (IEC/UL) Approved, File #255632**
- \blacksquare **3V to 5.5V Supply Operation**
- **Transformer Driver with Integrated Isolated Side LDO**
- ⁿ **SPI/Digital (LTM2810-S) or I2C (LTM2810-I) Options**
- ⁿ **High Common Mode Transient Immunity: 50kV/μs**
- \blacksquare High Speed Operation:
	- ⁿ **10MHz Digital Isolation**
	- ⁿ **4MHz/8MHz SPI Isolation**
	- ⁿ **400kHz I2C Isolation**
- Operation Up to 125°C (H-Grade)
- 1.62V to 5.5V Logic Supplies for Flexible Digital Interfacing
- \blacksquare \pm 25kV ESD HBM Across the Isolation Barrier
- **Naximum Continuous Working Voltage: 1kV**_{RMS}, 1.6kV_{DC}
- Low Current Shutdown Mode (<10µA)
- 22mm \times 6.25mm \times 2.06mm BGA Package

APPLICATIONS

- **EV/HEV Systems**
- Industrial and Metering Systems
- Test and Measurement Equipment
- Medical Equipment

TYPICAL APPLICATION

FEATURES DESCRIPTION

The LTM[®]2810 is a complete galvanic digital μ Module[®] (micromodule) isolator. No external components are required. Individual 3V to 5.5V supplies power each side of the digital isolator. Separate logic supply pins allow easy interfacing with different logic levels from 1.62V to 5.5V, independent of the main supply.

Module options are available with compatibility to SPI (LTM2810-S) and 1^2C (LTM2810-I), master mode only, specifications.

The module includes an integrated transformer driver on the logic side and an LDO on the isolated side to regulate the rectified transformer output. The LDO output is nominally 5V but may be overdriven.

Coupled inductors provide $7500V_{RMS}$ of isolation between the input and output logic interface. This device is ideal for systems with different ground potentials, allowing uninterrupted communication through large common mode transients faster than 50kV/μs.

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LTM2810-I Operating Through 70kV/µs CM Transients

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ABSOLUTE MAXIMUM RATINGS **(Note 1)**

PIN CONFIGURATION

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ORDER INFORMATION

• Device temperature grade is indicated by a label on the shipping container. • This product is not recommended for second side reflow.

This product is moisture sensitive. For more information, go to Recommended BGA PCB Assembly and Manufacturing Procedures.

• BGA Package and Tray Drawings

• Pad or ball finish code is per IPC/JEDEC J-STD-609.

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating

temperature range, otherwise specifications are at TA = 25°C. PVCC = VCC = 5V, VL = VL2 = 3.3V, VIN = GND = GND2 = 0V, ON = VL, and ON2 = VL2 unless otherwise noted. Specifications apply to all options unless otherwise noted.

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ISOLATION CHARACTERISTICS T_A = 25°C.

REGULATORY INFORMATION

CSA (Note 8)

CSA 60950-1-07+A1+A2 and IEC 60950-1, second edition, +A1 +A2: Basic Insulation at 1600V_{RMS} Reinforced Insulation at 800V_{RMS}

CSA 62368-1-14 and IEC 62368-1-14:2014, second edition:

Basic Insulation at 1600V_{RMS} Reinforced Insulation at 800V_{RMS}

- CSA 60601-1:14 and IEC 60601-1, third edition, +A1: Two means of patient protection (2 MOPP) at $500V_{RMS}$
- UL 1577-2015: Single Protection, 7500V_{RMS} Isolation Voltage

File 255632

ELECTRICAL CHARACTERISTICS

Note 1: Stresses beyond those listed under [Absolute Maximum Ratings](#page-1-0) may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: Guaranteed by design and not subject to production test.

Note 3: Guaranteed by other measured parameters and is not tested directly.

Note 4: This µModule isolator includes overtemperature protection that is intended to protect the device during momentary overload conditions. Junction temperature will exceed 125°C when overtemperature protection is active. Continuous operation above specified maximum operating junction temperature may result in device degradation or failure. **Note 5:** Device is considered a 2-terminal device. Pin group A1 through C6 shorted together and pin group V1 through X6 shorted together.

Note 6: The rated dielectric insulation voltage should not be interpreted as a continuous voltage rating.

Note 7: Maximum junction temperature limits operating conditions. The regulated output voltage specification does not apply for all possible combinations of input voltage and output current. Limit the output current range if operating at the maximum input-to-output voltage differential. Limit the input-to-output voltage differential if operating at maximum output current. Current limit foldback will limit the maximum output current as a function of input-to-output voltage.

Note 8: Ratings are for pollution degree 2, material group 3 and overvoltage category II where applicable. Ratings for other environmental and electrical conditions to be determined from the appropriate safety standard.

TYPICAL PERFORMANCE CHARACTERISTICS **TA = 25°C, PVCC = VCC = 5V, VL = VL2 = 3.3V,**

 $GND = GND2 = 0V$, $ON = V_L$, and $ON = V_{L2}$, unless otherwise noted.

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TYPICAL PERFORMANCE CHARACTERISTICS **TA ⁼ 25°C, PVCC = VCC ⁼ 5V, VL = VL2 ⁼ 3.3V,**

GND = GND2 = 0V, ON = VL, and ON2 = VL2, unless otherwise noted.

0

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CURRENT LIMIT (mA) 500 540 580 620 660 700 740 780 820 860 900

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PIN FUNCTIONS

LTM2810-S, Logic Side

DI3 (A1): Digital Input, Referenced to V_L and GND. Logic input connected to O3 through the isolation barrier. The logic state on DI3 translates to the same logic state on O3. Connect to GND or V_L if not used.

DI2 (A2): Digital Input, Referenced to V_L and GND. Logic input connected to O2N and O2P through the isolation barrier. The logic state on DI2 translates to the same logic state on O2N and O2P. Connect to GND or V_L if not used.

DI1 (A3): Digital Input, Referenced to V_1 and GND. Logic input connected to O1 through the isolation barrier. The logic state on DI1 translates to the same logic state on O1. Connect to GND or V_1 if not used.

ON (A4): Enable, Referenced to V_L and GND. Enables data communication through the isolation barrier. If ON is high the part is enabled and communications are functional to the isolated side. If ON is low the logic side is held in reset, all digital outputs are in a high impedance state. Connect to V_1 if not driven.

V_L (A5): Logic Supply. Interface supply voltage for pins DI1, DI2, DI3, DO1, DO2P, DO3, DO1E, and ON. Operating voltage is 1.62V to 5.5V. Internally bypassed with 1μF.

ST1, ST2 (A6, C6): Bridge Driver Outputs, Referenced to PV $_{CC}$ and GND. Each output runs at 50% duty cycle, ST1 is 180 degrees out of phase with ST2. Operating frequency is 2MHz. Bridge driver is enabled when PV_{CC} is between 3V to 5.5V and ON is high.

DO3 (B1): Digital Output, Referenced to V_L and GND. Logic output connected to I3 through the isolation barrier. Under the condition of an isolation communication failure this output is in a high impedance state.

DO2N (B2): Open Drain Pull-Down Output to GND. Logic output connected to I2 through the isolation barrier. Under the condition of an isolation communication failure this output is in a high impedance state.

DO2P (B3): Open Drain Pull-Up Output to V_L. Logic output connected to I2 through the isolation barrier. Under the condition of an isolation communication failure this output is in a high impedance state.

DO1 (B4): Digital Output, Referenced to V_L and GND. Logic output connected to I1 through the isolation barrier. Under the condition of an isolation communication failure this output is in a high impedance state.

PV_{CC} (B5): Bridge Driver Supply Voltage. Operating voltage is 3V to 5.5V, connect to GND to disable bridge driver. Internally bypassed with 2.2μF.

DO1E (C4): Digital Output Enable, Referenced to V_L and GND. A logic high on DO1E places the logic side DO1 pin in a high impedance state, a logic low enables the output. Connect to GND or V_1 if not used.

V_{CC} (C5): Supply Voltage. Operating voltage is 3V to 5.5V. Internally bypassed with 1μF.

GND (B6, C1 to C3): Circuit Ground.

LTM2810-S, Isolated Side

O1E (V4): Digital Output Enable, Referenced to V_{L2} and GND2. A logic high on $\overline{O1E}$ places the isolated side O1 pin in a high impedance state, a logic low enables the output. Connect to GND2 or V_1 ₂ if not used.

V_{IN} (V6): Internal LDO Input Voltage. Operating voltage is 3.6V to 38V. Internally bypassed with 0.1μF.

03 (W1): Digital Output, Referenced to V_{L2} and GND2. Logic output connected to DI3 through the isolation barrier. Under the condition of an isolation communication failure O3 defaults to a high state.

O2N (W2): Open Drain Pull-Down Output to GND2. Logic output connected to DI2 through the isolation barrier. Under the condition of an isolation communication failure O2N defaults to a low state.

O2P (W3): Open Drain Pull-Up Output to V_{L2}. Logic output connected to DI2 through the isolation barrier. Under the condition of an isolation communication failure O2P defaults to a low state.

O1 (W4): Digital Output, Referenced to V_{12} and GND2. Logic output connected to DI1 through the isolation barrier. Under the condition of an isolation communication failure O1 defaults to a low state.

AVL2 (W5): VL2 LDO Adjust Pin.

PIN FUNCTIONS

I3 (X1): Digital Input, Referenced to V_{L2} and GND2. Logic input connected to DO3 through the isolation barrier. The logic state on I3 translates to the same logic state on DO3. Connect to GND2 or V_{L2} if not used.

I2 (X2): Digital Input, Referenced to V_{L2} and GND2. Logic input connected to DO2N and DO2P through the isolation barrier. The logic state on I2 translates to the same logic state on DO2N and DO2P. Connect to GND2 or V_{L2} if not used.

I1 (X3): Digital Input, Referenced to V_{L2} and GND2. Logic input connected to DO1 through the isolation barrier. The logic state on I1 translates to the same logic state on DO1. Connect to GND2 or V_{L2} if not used.

ON2 (X4): Enable, Referenced to V_{L2} and GND2. Enables data communication through the isolation barrier. If ON2 is high the part is enabled and communications are functional to the logic side. If ON2 is low the isolated side is held in reset, O1, O2N and O2P are in a low state, and O3 is in a high state. Connect to V_{L2} if not driven.

V₁₂ (X5, X6): Logic Supply, Referred to GND2. Interface supply voltage for pins $01, 02$ P, $03, 11, 12, 13, 01$ E, and ON2. Operating voltage is 3V to 5.5V. Internally bypassed with 6.6μF.

GND2 (V1 to V3, V5, W6): Isolated Ground.

LTM2810-I, Logic Side

DI3 (A1): Digital Input, Referenced to V_1 and GND. Logic input connected to O3 through the isolation barrier. The logic state on DI3 translates to the same logic state on O3. Connect to GND or V_1 if not used.

SDA (A2, B2): Serial I²C Data Pins, Referenced to V_L and GND. Bidirectional logic pins connected to isolated side SDA2 pins through the isolation barrier. Under the condition of an isolation communication failure pins are in a high impedance state. Pull up to V_L if not used.

SCL (A3): Serial I²C Clock Input, Referenced to V_L and GND. Logic input connected to isolated side SCL2 pin through the isolation barrier. Clock is unidirectional from logic to isolated side. Pull up to V_L if not used.

ON (A4): Enable, Referenced to V_L and GND. Enables data communication through the isolation barrier. If ON is high the part is enabled and communications are functional to the isolated side. If ON is low the logic side is held in reset, all digital outputs are in a high impedance state. Connect to V_1 if not driven.

V_L (A5): Logic Supply. Interface supply voltage for pins SCL, DI3, DO1, DO3, and ON. Operating voltage is 3V to 5.5V. Internally bypassed with 1μF.

ST1, ST2 (A6, C6): Bridge Driver Outputs, Referenced to PV $_{\text{CC}}$ and GND. Each output runs at 50% duty cycle, ST1 is 180 degrees out of phase with ST2. Operating frequency is 2MHz. Bridge driver is enabled when PV_{CC} is between 3V to 5.5V and ON is high.

DO3 (B1): Digital Output, Referenced to V_L and GND. Logic output connected to I3 through the isolation barrier. Under the condition of an isolation communication failure this output is in a high impedance state.

DO1 (B4): Digital Output, Referenced to V_L and GND. Logic output connected to I1 through the isolation barrier. Under the condition of an isolation communication failure this output is in a high impedance state.

PV_{CC} (B5): Bridge Driver Supply Voltage. Operating voltage is 3V to 5.5V, connect to GND to disable bridge driver. Internally bypassed with 2.2μF.

V_{CC} (C5): Supply Voltage. Operating voltage is 3V to 5.5V. Internally bypassed with 1μF.

GND (B3, B6, C1 to C4): Circuit Ground.

LTM2810-I, Isolated Side

V_{IN} (V6): Internal LDO Input Voltage. Operating voltage is 3.6V to 38V. Bypassed with 0.1μF.

03 (W1): Digital Output, Referenced to V_{L2} and GND2. Logic output connected to DI3 through the isolation barrier. Under the condition of an isolation communication failure O3 defaults to a high state.

SDA2 (W2, W3, X2): Serial ²C Data Pins, Referenced to V_1 ₂ and GND2. Bidirectional logic pins connected to logic side SDA pins through the isolation barrier. Output is biased high by a 1.8mA current source. Do not connect

PIN FUNCTIONS

an external pull-up device to SDA2. Under the condition of an isolation communication failure outputs default to a high state. Pins connected internally.

SCL2 (W4): Serial I²C Clock Output, Referenced to V_{L2} and GND2. Logic output connected to logic side SCL pin through the isolation barrier. Clock is unidirectional from logic to isolated side. SCL2 has a push-pull output stage; do not connect an external pull-up device. Under the condition of an isolation communication failure this output defaults to a high state.

AV_{L2} (W5): V_{L2} LDO Adjust Pin.

I3 (X1): Digital Input, Referenced to V_{L2} and GND2. Logic input connected to DO3 through the isolation barrier. The logic state on I3 translates to the same logic state on DO3. Connect to GND2 or V_{12} if not used.

I1 (X3): Digital Input, Referenced to V_{12} and GND2. Logic input connected to DO1 through the isolation barrier. The logic state on I1 translates to the same logic state on DO1. Connect to GND2 or V_1 ₂ if not used.

ON2 (X4): Enable, Referenced to V_{L2} and GND2. Enables data communication through the isolation barrier. If ON2 is high the part is enabled and communications are functional to the logic side. If ON2 is low the isolated side is held in reset, all digital outputs are in a high state. Connect to V_1 ₂ if not driven.

V₁₂ (X5, X6): Logic Supply, Referred to GND2. Interface supply voltage for pins SCL2, SDA2, 11, 13, 03, and ON2. Operating voltage is 3V to 5.5V. Internally bypassed with 6.6μF.

GND2 (W6, V1 to V5): Isolated Ground.

BLOCK DIAGRAM

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BLOCK DIAGRAM

TEST CIRCUIT

Figure 1. Logic Timing Measurements

Figure 2. Logic Enable/Disable Time

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TEST CIRCUIT

Figure 4. ONx Enable/Disable Time

Overview

The LTM2810 digital µModule isolator provides a galvanically-isolated robust logic interface, complete with decoupling capacitors. The LTM2810 is ideal for use in networks where grounds can take on different voltages. Isolation in the LTM2810 blocks high voltage differences and eliminates ground loops, and is extremely tolerant of common mode transients between ground planes. Errorfree operation is maintained through common mode events as fast as 50kV/μs providing excellent noise isolation.

Input Supply (V_{CC})

The LTM2810 is powered by a 3V to 5.5V supply on the logic side of the isolation interface. The input supply provides power to the internal isolated communications interface and is completely independent of either the logic power supply or bridge driver supply. V_{CC} is bypassed internally with a 1µF ceramic capacitor.

Bridge Driver Supply (PV_{CC})

The integrated bridge driver is powered by a 3V to 5.5V supply on the logic side of the isolation interface. The bridge driver may be disabled by tying PV_{CC} to GND. PV_{CC} is bypassed internally with a 2.2µF ceramic capacitor.

LDO Input Supply (V_{IN})

The isolated side includes an integrated LDO powered by V_{IN} , with a nominal output voltage of 5V on V_{12} . Input operating range is 4V to 38V. V_{IN} is bypassed internally with a 0.1μ F ceramic capacitor. V_{IN} may be grounded or left unconnected if $V_{1,2}$ is driven by an external supply.

Logic Supplies (V_L, V_{L2})

Separate logic supply pins, V_L and V_{L2} , allow the LTM2810 to interface with any logic signal from 1.62V to 5.5V on the logic side of the SPI/Digital version, 3V to 5.5V for the logic and isolated sides of the I2C version and isolated side of the SPI/Digital version, as shown in [Figure 5.](#page-14-0)

Figure 5. Supplies are Independent

With V_{IN} driven, the V_{L2} output may be adjusted from the nominal 5V by connecting a voltage divider to the AV_{12} pin as shown in [Figure 5](#page-14-0). Select the voltage divider ratio so that the AV_{L2} voltage is 0.6V. The value of R1 should be no greater than 13k Ω to minimize errors in the output voltage caused by the adjust pin bias current and internal voltage divider.

There is no interdependency between V_{CC} , V_L , and PV_{CC} . They may simultaneously operate at any voltage within their specified operating ranges and may sequence in any order. V_1 is bypassed internally with a 1µF ceramic capacitor and V_{L2} is bypassed internally by a 6.6 μ F ceramic capacitor.

Hot Plugging Safely

Caution must be exercised in applications where power is plugged into the LTM2810's power supplies, V_{CC} , PV $_{CC}$, V_1 , or V_1 ₂ due to the integrated ceramic decoupling capacitors. The parasitic cable inductance along with the high-Q characteristics of ceramic capacitors can cause substantial ringing which could exceed the maximum voltage ratings and damage the LTM2810. Refer to [Application Note 88](https://www.analog.com/media/en/technical-documentation/application-notes/an88f.pdf?doc=LTM2810.pdf), entitled Ceramic Input Capacitors Can Cause Overvoltage Transients for a detailed discussion and mitigation of this phenomenon.

Isolation Transformer

Because of the wide voltage range for V_{IN} , there are many options for the isolation transformer and rectifier topology. The selected transformer should meet the application's isolation requirements, have a minimum volt-second rating greater than $0.5 \cdot PV_{CC}$ µVS and a current rating sufficient for the LTM2810's V_{IN} plus any application load. Different winding configurations and turns ratios allow the accommodation of different input or isolated output voltages as illustrated by [Figures 14 to 18.](#page-24-0)

Channel Timing Uncertainty

Multiple channels are supported across the isolation boundary by encoding and decoding of the inputs and outputs. Up to three signals in each direction are assembled as serial packets and transferred across the isolation barrier. The time required to transfer all three bits is 100ns maximum, and sets the limit for how often a signal can change on the opposite side of the barrier. Encoding and transmission is independent for each data direction. The technique used assigns DI1(-S) or SCL(-I) on the logic side, and I1(-S or -I) on the isolated side, the highest priority such that there is no jitter on the associated output channels, only delay. This preemptive scheme will produce a certain amount of uncertainty on the other isolation channels. The resulting pulse width uncertainty on these low priority channels is typically ± 6 ns, but may vary up to ± 44 ns if the low priority channels are not encoded within the same high priority serial packet.

Serial Peripheral Interface (SPI) Bus

The LTM2810-S provides an SPI compatible isolated interface. The maximum data rate is a function of the inherent channel propagation delays, channel to channel pulse width uncertainty, and data direction requirements. Channel timing is detailed in [Figures 6 through 9,](#page-17-0) [Table 2](#page-18-0), and [Table 3](#page-20-0). The SPI protocol supports four unique timing configurations defined by the clock polarity (CPOL) and clock phase (CPHA) summarized in [Table 1.](#page-15-0)

Table 1. SPI Mode

The maximum data rate for bidirectional communication is 4MHz, based on a synchronous system, as detailed in the timing waveforms. Slightly higher data rates may be achieved by skewing the clock duty cycle and minimizing the DO1 (SDO) to DI1 (SCK) setup time, however the clock rate is still dominated by the system propagation delays. A discussion of the critical timing paths relative to [Figure 6](#page-17-1) and [Figure 7](#page-17-2) follows. For SPI communication DI1 = SCK, $D12 = SDI$, $D13 = CS$, $D01 = SDO$, $D1 = SCK2$, $O2N$ and $O2P = SD12$, $O3 = CS2$, and $11 = SD02$.

- CS to SCK (master sample SDO, 1st SDO valid)
	- $t_0 \rightarrow t_1$ \approx 50ns, CS to CS2 propagation delay
	- $t_1 \rightarrow t_1$ + Isolated slave device propagation (response time), asserts SDO2
	- $t_1 \rightarrow t_3$ \approx 50ns, SDO2 to SDO propagation delay
	- $t_3 \rightarrow t_5$ Setup time for master SDO to SCK
- SDI to SCK (master data write to slave)
	- $t_2 \rightarrow t_4$ \approx 50ns, SDI to SDI2 propagation delay
	- $t_5 \rightarrow t_6$ \approx 50ns, SCK to SCK2 propagation delay
	- $t_2 \rightarrow t_5$ ≥ 50 ns, SDI to SCK, separate packet non-zero setup time
	- $t_4 \rightarrow t_6$ \geq 50ns, SDI2 to SCK2, separate packet nonzero setup time
- SDO to SCK (master sample SDO, subsequent SDO valid)
	- t₈ Setup data transition SDI and SCK
	- $t_8 \rightarrow t_{10}$ ≈ 50ns, SDI to SDI2 and SCK to SCK2 propagation delay
	- t_{10} SDO2 data transition in response to SCK2
	- $t_{10} \rightarrow t_{11} \approx 50$ ns, SDO2 to SDO propagation delay

 $t_{11} \rightarrow t_{12}$ Setup time for master SDO to SCK

Maximum data rate for single direction communication, master to slave, is 8MHz, limited by the systems encoding/decoding scheme or propagation delay. Timing details for both variations of clock phase are shown in [Figure 8](#page-19-0), [Figure 9](#page-19-1), and [Table 3](#page-20-0).

Additional requirements to insure maximum data rate are:

- CS is transmitted prior to (asynchronous) or within the same (synchronous) data packet as SDI
- SDI and SCK setup data transition occur within the same data packet. Referencing [Figure 6](#page-17-1), SDI can precede SCK by up to 13ns ($t_7 \rightarrow t_8$) or lag SCK by 3ns ($t_8 \rightarrow t_9$) and not violate this requirement. Similarly in [Figure 8](#page-19-0), SDI can precede SCK by up to 13ns $(t_4 \rightarrow t_5)$ or lag SCK by 3ns ($t_5 \rightarrow t_6$).

Table 2. Bidirectional SPI Timing Event Description

Table 3. Unidirectional SPI Timing Event Description

Inter-IC Communication (I2C) Bus

The LTM2810-I provides an isolated $1²C$ compatible interface supporting master mode only, with a unidirectional clock (SCL), and bidirectional data (SDA). The maximum data rate is 400kHz which supports fast-mode $1²C$. Timing is detailed in [Figure 10.](#page-21-0) The data rate is limited by the slave acknowledge setup time (t_{SUEACK}), consisting of the I²C standard minimum setup time (t_{SULDAT}) of 100ns, maximum clock propagation delay of 225ns, glitch filter and isolated data delay of 500ns maximum, and the combined isolated and logic data fall time of 300ns at maximum bus loading. The total setup time reduces the $1²C$ data hold time $(t_{HD; DAT})$ to a maximum of 175ns, guaranteeing sufficient data setup time (t_{SULACK}) .

The isolated side bidirectional serial data pin, SDA2, simplified schematic is shown in [Figure 11](#page-21-1). An internal 1.8mA current source provides a pull-up for SDA2. Do not connect any other pull-up device to SDA2. This current source is sufficient to satisfy the system requirements for bus capacitances greater than 200pF in fast mode and greater than 400pF in standard mode.

Additional proprietary circuitry monitors the slew rate on the SDA and SDA2 signals to manage directional control across the isolation barrier. Slew rates on both pins must be greater than 1V/us for proper operation.

Figure 10. I 2C Timing Diagram

Figure 11. Isolated SDA2 Pin Schematic

The logic side bidirectional serial data pin, SDA, requires a pull-up resistor or current source connected to V_L . Follow the requirements in [Figure 12](#page-22-0) and [Figure 13](#page-22-1) for the appropriate pull-up resistor on SDA that satisfies the desired rise time specifications and V_{OL} maximum limits for fast and standard modes. The resistance curves represent the maximum resistance boundary; any value may be used to the left of the appropriate curve.

The isolated side clock pin, SCL2, has a weak push-pull output driver; do not connect an external pull-up device. SCL2 is compatible with $1²C$ devices without clock stretching. On lightly loaded connections, a 100pF capacitor from SCL2 to GND2 or RC low pass filter (R = 500Ω , $C = 100pF$) can be used to decrease the rise and fall times and minimize noise.

Some consideration must be given to signal coupling between SCL2 and SDA2. Separate these signals on a printed circuit board or route with ground between. If these signals are wired off board, twist SCL2 with V_{12} and/or GND2 and SDA2 with GND2 and/or V_L ; do not twist SCL2 and SDA2 together. If coupling between SCL2 and SDA2 is unavoidable, place the aforementioned RC filter at the SCL2 pin to reduce noise injection onto SDA2.

Figure 12. Maximum Standard Speed Pull-Up Resistance on SDA

Figure 13. Maximum Fast Speed Pull-Up Resistance on SDA

RF, Magnetic Field Immunity

The isolator µModule technology used within the LTM2810 has been independently evaluated, and successfully passed the RF and magnetic field immunity testing requirements per European Standard EN 55024, in accordance with the following test standards:

- EN 61000-4-3 Radiated, radio-frequency, electromagnetic field immunity
- EN 61000-4-8 Power frequency magnetic field immunity
- EN 61000-4-9 Pulsed magnetic field immunity

Tests were performed using an unshielded test card designed per the data sheet PCB layout recommendations. Specific limits per test are detailed in [Table 4.](#page-23-0)

Table 4. EMC Immunity Tests

* Non IEC method.

PCB Layout

The high integration of the LTM2810 makes PCB layout very simple. However, to optimize its electrical isolation characteristics and EMI performance, some layout considerations are necessary.

• Input and output supply decoupling is not required, since these components are integrated within the package. An additional bulk capacitor with a value of 6.8µF to 22µF with 1 Ω to 3 Ω of ESR is recommended. The high ESR of this capacitor reduces board resonances and minimizes voltage spikes caused by hot plugging of the supply voltage. For EMI sensitive applications, an additional low ESL ceramic capacitor of 1µF to 4.7µF,

placed as close to the power and ground terminals as possible, is recommended. Alternatively, a number of smaller value parallel capacitors may be used to reduce ESL and achieve the same net capacitance.

- Do not place copper on the PCB between the inner columns of pads. This area must remain open to withstand the rated isolation voltage.
- The use of solid ground planes for GND and GND2 is recommended for non-EMI critical applications to optimize signal fidelity, and minimize RF emissions due to uncoupled PCB trace conduction. The drawback of using ground planes where EMI is of concern, is the creation of a dipole antenna structure which can radiate differential voltages formed between GND and GND2. If ground planes are used, it is recommended to minimize their area, and use contiguous planes as any openings or splits can exacerbate RF emissions.
- For large ground planes a small capacitance (≤ 330pF) from GND to GND2, either discrete or embedded within the substrate, provides a low impedance current return path for the module parasitic capacitance, minimizing any high frequency differential voltages and substantially reducing radiated emissions. Discrete capacitance will not be as effective due to parasitic ESL. In addition, voltage rating, leakage, and clearance must be considered for component selection. Embedding the capacitance within the PCB substrate provides a near ideal capacitor and eliminates component selection issues; however, the PCB must be 4 layers. Care must be exercised in applying either technique to ensure the voltage rating of the barrier is not compromised.
- In applications without an embedded PCB substrate capacitance, a slot may be added between the logic side and isolated side device pins. The slot extends the creepage path between terminals on the PCB side, and may reduce leakage caused by PCB contamination. The slot should be placed in the middle of the device and extend beyond the package perimeter.

TYPICAL APPLICATIONS

Figure 14. Isolated SPI with 5V Input and 5V Regulated Output

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TYPICAL APPLICATIONS

Figure 16. Isolated SPI with 3.3V Input and 5V Regulated Output

Figure 17. Isolated SPI with 3.3V Input with 3.3V and 5V Regulated Outputs

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TYPICAL APPLICATIONS

*NOTE: I_{MON} VALID WHEN INP (GATE) IS HIGH FOLLOWED BY 150µs OF BLANKING TIME. TGUP – TS VOLTAGE IS APPROXIMATELY 7V.

Figure 18. High Voltage Switch Controller with Current Readback

PACKAGE DESCRIPTION

REVISION HISTORY

Rev. A

TYPICAL APPLICATION

Figure 19. High Voltage 30A Power/Energy Monitor with Integrated Sense Resistor

RELATED PARTS

Rev. A